

GaN HEMTs are Still Ongoing

Kazutaka Takagi

Microwave Solid-state Engineering Dept.,
Komukai Complex, Toshiba Corporation

Abstract — Many new GaN/AlGaIn HEMTs products have begun to bloom. On the other hand, the development and improvement of GaN/AlGaIn HEMTs has continued. In this paper, we show the current status of GaN HEMTs. We also show two of our new works. The first one is an optimization of a source-field-plate for harmonics control. The second one is an MMICs in Ka-band.

Index Terms — GaN HEMTs, efficiency, harmonics. Ka-band

I. INTRODUCTION

The first DC performance of AlGaIn/GaN HEMT was shown in 1993 [1] and the first RF performance was demonstrated in 1996 with 1.1 W/mm of the output power density with 18.6 % of Power-Added-Efficiency (PAE) at 2 GHz for an AlGaIn/GaN HEMT [2].

Many AlGaIn/GaN HEMTs showed a discrepancy between the predicted output powers from static I-V curves because of a current-collapse that was a trap-related phenomenon to which both surface and bulk traps contributed. The current-collapse made the output power and PAE low. Some field-plates were adopted in order to compensate for the quality of epitaxial wafers and to reduce the current-collapse [22]-[24].

The output power density increased with the help of steadily improved growth techniques, material qualities, enhanced processing technologies and more optimum device designs including a gate-field-plate and source-field-plate structures. Then, a 32 W/mm in 2004 [3] and 41 W/mm in 2006 [4] were reported.

The high power density needed a high efficiency for reducing the generation of heat. The PAE also increased in each frequency. 85% of PAE with 16.5 W of output power at 2 GHz in 2006 [5], 80% of PAE with 4 W of output power at 4 GHz in 2010 [6], 79% of PAE with 2.1 W of output power at 5.65 GHz in 2012 [7], over 70% of PAE with 10 W of output power at 10 GHz in 2012 [8], 50% of PAE with 3.7 W of output power at 15 GHz in 2011 [9], and 52% of PAE with 1.2 W of output power at 30 GHz in 2013 [10] were reported.

After that, many products for radar systems were released. Products with 50% to 65% of PAE with over 300 W of output power were released at S-band [11], [12]. Products with 40% to 54% of PAE with over 80 W of output power were released at X-band [14], [15].

Many products for satellite communication systems were also released. Products with around 50% of PAE with over 100 W of output power were released at C-band [16]. Products with around 40% of PAE with over 100 W of output power were released at X-band [17], [18]. Products with around 30%

of PAE with over 50 W of output power were released at Ku-band [19] from several companies. Products with around 30% of PAE with around 10 W of output power were released at Ka-band [20], [21].

On the other hand, the development and improvement of GaN/AlGaIn HEMTs continued. Now the quality of epitaxial wafers has become much better. And harmonics controls have been adopted. Next, it was determined that the optimum device designs should be changed.

Interest has also moved to higher frequencies. Here, we show two of our new works. The first one is an optimization of a source-field-plate for harmonics control. The second one is an MMICs in Ka-band.

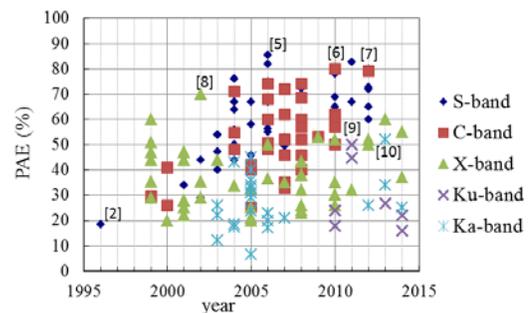


Fig. 1. PAE of AlGaIn/GaN HEMTs versus year.

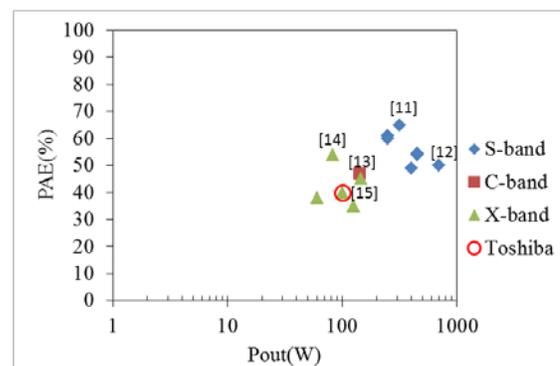


Fig. 2. PAE of recent AlGaIn/GaN HEMTs products for radar systems versus Pout.

II. ALGAN/GAN HEMTs WITH FIELD-PLATE

Field-plates were necessary for reducing the current collapse, but they also decreased the gain by increasing the parasitic capacitance.

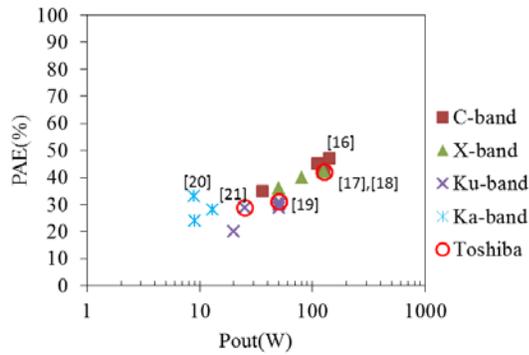


Fig. 3. PAE of recent AlGaN/GaN HEMTs products for satellite communication systems versus Pout.

Two kinds of AlGaN/GaN epitaxial wafers were used in this study, called “Epi wafer A” and “Epi wafer B” [25]. The wafers were grown on SI SiC substrates by MOCVD with a difference in the GaN buffer layer. “Epi wafer B” has considerably lower-density trap states than “Epi wafer A”.

Three device structures were then fabricated on both wafers. They were (a) without a source field plate, (b) with a source field plate $L_{sfp}=0.5\mu\text{m}$ and (c) $L_{sfp}=1.5\mu\text{m}$. All devices had the same gate length (L_g) of $0.6\mu\text{m}$ and a total gate width of $540\mu\text{m}$ ($270\mu\text{m} \times 2$).

Fig.4 shows the drain efficiencies of the devices when the devices were biased at 40V drain supply at a fundamental frequency of 3.0 GHz on a harmonic tuning active loadpull system.

When the harmonics were shorted, each device showed almost the same drain efficiency. But when the harmonics were optimized, each device showed a difference.

For devices fabricated on “Epi wafer A”, the drain efficiency increased with the longer source field plate. In contrast, for devices fabricated on “Epi wafer B”, the drain efficiency decreased with the longer source field plate. And the drain efficiencies of “Epi wafer B” were higher than those of “Epi wafer A”. The device without a source field plate on “Epi wafer B” showed a drain efficiency of 84% which was about 20% higher than that of “Epi wafer A”. To achieve high efficiency, the harmonics control needed a gain at the harmonics frequencies which were two or three times higher than the fundamental frequency.

Two 35mm-gate-width dies which had a source field plate of $0.5\mu\text{m}$ were attached with internal matching circuits into a conventional copper package. Fig.5 shows the power characteristics under the pulsed condition of 200 μs pulse width and 10% duty cycles. The measured output power reached 461W (56.64dBm) with a 15.44dB related gain and 59.4% of PAE at a drain voltage of 50V.

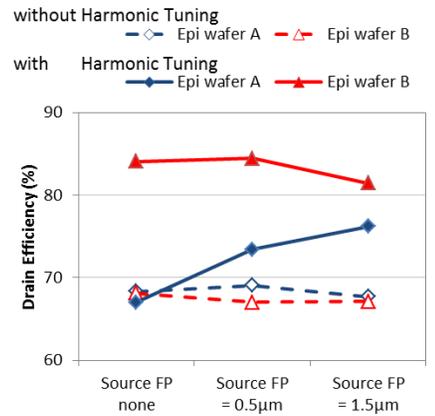


Fig. 4. Dependency of the drain efficiency on the epitaxial wafer and the device structure. The devices were biased at $V_{ds}=40\text{ V}$.

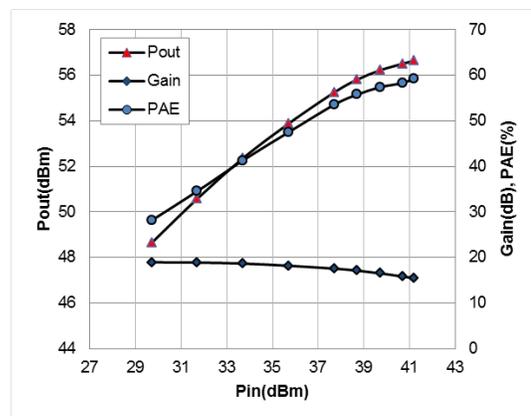


Fig. 5. A device with two 35mm-gate-width dies which had a source field plate of $0.5\mu\text{m}$ produced output power reached 461W (56.64dBm) with a 15.44dB related gain and 59.4% of PAE at a drain voltage of 50V under the pulsed condition of 200 μs pulse width and 10% duty cycles.

III. ALGAN/GAN HEMTs WITHOUT FIELD-PLATED

At a higher frequency, since it is difficult to realize a high gain, structures which do not drop the gain are needed. In order for the GaN devices to achieve optimal performance at millimeter-wave frequencies, special attention was paid to the formation of the epitaxial layer structure, the gate fabrication techniques and the layout pattern.

A $50\mu\text{m}$ gate-width 8-finger cell was designed for the frequency. A Loadpull measurement was performed on the FET cell to determine the optimum load impedance using a passive-tuner system at 31GHz [26]. The cell achieved a saturated output power of 31.2dBm (3.3W/mm), an associated gain of 7dB and PAE of 40%.

A 2-stage MMIC was designed with 16 cells for the output power stage (6.4mm) which was driven by 8 FET cells for the driver stage (3.2mm). At $V_{DD}=28\text{V}$ under pulse condition of 100 μs pulse width and 10% duty cycles, the MMIC achieved a saturated output power of 42.8dBm (19W) to 43.3dBm (21W) across 29GHz to 31GHz as shown in Fig. 6.

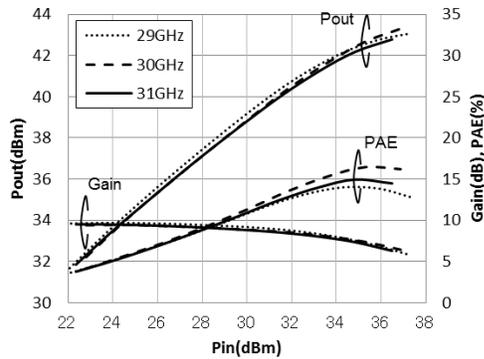


Fig. 6. MMIC 16 cells for an output power stage (6.4mm) which was driven by 8 FET cells for the driver stage (3.2mm) achieved a saturated output power of 42.8dBm (19W) to 43.3dBm (21W) across 29GHz to 31GHz at VDD=28V under pulse condition of 100 μ s pulse width and 10% duty cycles.

IV. CONCLUSION

The device without a source field plate on “Epi wafer B” which had a low-density trap state showed a drain efficiency of 84% at 3.0 GHz. Two 35mm-gate-width dies which had a source field plate of 0.5 μ m achieved 461W (56.64dBm) with 15.44dB related gain and 59.4% of PAE at a drain voltage of 50V.

A 50 μ m gate-width 8-finger cell showed a saturated output power of 31.2dBm (3.3W/mm), an associated gain of 7dB and PAE of 40%. A 2-stage MMIC achieved a saturated output power of 42.8dBm (19W) to 43.3dBm (21W) across 29GHz to 31GHz.

In order to realize high PAE, harmonics control is effective. The harmonics control requires a high breakdown voltage and a high gain of devices, so the GaN HEMT is the best device for high power and high efficiency.

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